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U.S. PATENT APPLICATION

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Invention: VEHICULAR CONTROL SYSTEM

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SPECIFICATION

VEHICULAR CONTROL SYSTEM

CROSS-REFERENCE TO RELATED APPLICATION

This application is based on Japanese Patent Application No.
5 2003-79368 filed March 24, 2003, the disclosure of which is
incorporated herein by reference.

FIELD OF THE INVENTION

The present invention relates to a vehicular control system
10 disposed with a feedforward control function.

BACKGROUND OF THE INVENTION

In relation to vehicular control systems, there are vehicular
control systems such as described in Japanese Patent No. 3316955
15 where a controlled system is modeled, a model constant is calculated
in real time, a feedback gain is calculated on the basis of the
model constant, and a controlled value of a controlled system is
made to follow a target value to conduct feedback control.

However, because an error between the target value and the
20 actual controlled value is generated and the feedback control works
to reduce this error, there has been the drawback that responsiveness
is relatively slow.

Thus, a control system configured to combine and execute
feedforward control, whose responsiveness is fast, with feedback
25 control has been developed.

However, because conventional feedforward control has been
configured to calculate a feedforward corrected value using a

predetermined gain, there has been the drawback that the effects of characteristic variations in the controlled system arising due to variations in the manufacture of the controlled system, temporal changes and changes in environmental conditions and operational conditions are not reflected in the feedforward corrected value, so that the control precision of feedforward control changes due to characteristic variations in the controlled system.

SUMMARY OF THE INVENTION

An object of the present invention is to provide a vehicular control system that can conduct feedforward control reflecting the effects resulting from characteristic variations in a controlled system and can execute highly responsive and highly precise feedforward control.

In order to achieve this object, the invention provides a vehicular control system that conducts feedforward control so that a controlled value of a controlled system disposed in a vehicle is made to follow a target value, the vehicular control system comprising: gain calculating means for adaptively determining a gain based on a value obtained by multiplying a derivative value of the target value by the error between the target value and the actual controlled value; and feedforward corrected value calculating means for determining, as a feedforward corrected value, a value obtained by multiplying the gain by the derivative value of the target value.

By configuring the vehicular control system in this manner, the gain can be automatically adjusted in accordance with

characteristic variations in the controlled system, feedforward control reflecting effects resulting from characteristic variations in the controlled system can be conducted, and the control precision of feedforward control can be improved.

5 Moreover, because a control equation that calculates the input (control input) of the controlled system from the target value serves as an inverse model of the transfer function of the controlled system, as will be described later, the output (controlled value) of the controlled system can be made to match the target value and highly
10 responsive feedforward control can be realized.

 In the present invention, because the derivative value of the target value used in calculating the feedforward corrected value becomes 0 in a steady state where the target value does not change, the effects of steady-state deviation between the target value and
15 the actual controlled value can be eliminated by multiplying the derivative value of the target value.

BRIEF DESCRIPTION OF THE DRAWINGS

 Fig. 1 is a diagram showing the schematic configuration of
20 an entire engine control system in a first embodiment of the invention;

 Fig. 2 is a block diagram describing a derivation method of a control expression used in the first embodiment;

 Fig. 3 is a flow chart showing the flow of processing of an electronic throttle control program of the first embodiment;

25 Figs. 4A and 4B are time charts describing an example of the electronic throttle control of the first embodiment;

 Fig. 5 is a block diagram describing an air-fuel ratio control

system of a second embodiment;

Fig. 6 is a flow chart showing the flow of processing of an air-fuel control program of the second embodiment;

Fig. 7 is a block diagram describing a derivation method of a control expression used in a third embodiment;

Fig. 8 is a flow chart showing the flow of processing of an electronic throttle control program of the third embodiment;

Fig. 9 is a flow chart showing the flow of processing of an air-fuel ratio control program of a fourth embodiment; and

Figs. 10A and 10B are time charts describing an example of the air-fuel ratio control of the fourth embodiment.

DESCRIPTION OF THE PREFERRED EMBODIMENTS

First Embodiment

A first embodiment where the invention is applied to an electronic throttle system will be described below on the basis of Figs. 1 to 4B.

A schematic configuration of an entire engine control system is described on the basis of Fig. 1. An air cleaner 13 is disposed at the most upstream portion of an intake pipe 12 of an engine 11, which is an internal combustion engine, and an air flow meter 14 that detects the intake air amount is disposed at a downstream side of the air cleaner 13. A throttle valve 15 whose opening is adjusted by a motor 17 such as a DC motor and a throttle opening sensor 16 that detects the throttle opening are disposed at a downstream side of the air flow meter 14.

A surge tank 12a is disposed at a downstream side of the throttle

valve 15, and an intake pipe pressure sensor 18 that detects the intake pipe pressure is disposed at the surge tank 12a. An intake manifold 19 that introduces air to each cylinder of the engine 11 is disposed at the surge tank 12a, and a fuel injection valve 20 that injects fuel is attached in the vicinity of an intake port of the intake manifold 19 of each cylinder. A spark plug 21 is attached to each cylinder at a cylinder head of the engine 11. Mixed air inside the pipes is combusted by the spark discharge of each spark plug 21.

10 A catalyst 23 such as a three-way catalyst that purifies CO, HC and NOx in exhaust gas is disposed at an exhaust pipe 22 of the engine 11, and an air-fuel ratio sensor 24 (or oxygen sensor) that detects the air-fuel ratio of the exhaust gas is disposed at an upstream side of the catalyst 23. A water temperature sensor 25 that detects the cooling water temperature and a crank angle sensor 26 that outputs a pulse signal each time a crankshaft of the engine 11 revolves by a constant crank angle (e.g., 30° CA) are disposed at a cylinder block of the engine 11. The crank angle and the engine revolving speed are detected on the basis of the output signal of
20 the crank angle sensor 26.

 The output of each sensor is inputted to an engine control circuit (represented below as "ECU") 27. The ECU 27 is mainly configured by a microcomputer and executes various engine control programs stored in an internally disposed ROM (Read Only Memory),
25 whereby the ECU 27 controls the fuel injection amount of the fuel injection valve 20 and the ignition timing of the spark plugs 21 depending on an operation state of the engine.

Moreover, the ECU 27 uses an electronic throttle system as feedforward control (represented below as "F/F control") and feedback control (represented below as "F/B control") to control the throttle opening to a target throttle opening set in accordance with an accelerator opening (accelerator control input) detected by an accelerator sensor (not shown). In this case, the ECU 27 executes a later-described electronic throttle control program of Fig. 3, whereby the ECU 27 corrects, by adaptive control, excess and deficiency with the F/F control and the F/B control.

A control used in the electronic throttle control program of Fig. 3 will be described below. In the first embodiment, as shown in Fig. 2, a controlled system (electronic throttle system) is approximated by a first-order lag system. In this case, an output y (actual throttle opening) of the controlled system can be made to match a target value y_m (target throttle opening) as long as the control (inverse model of transfer function of the controlled system) in the dotted lines of Fig. 2 can be realized.

However, because a time constant K of the controlled system is unknown or varies, it cannot be expressed with the control equation of Fig. 2.

Thus, in the first embodiment, a method that detects and controls the time constant K of the controlled system is adopted.

Here, the control equation of the transfer function is expressed by the following equation when the estimated value of the time constant K is represented by K_h .

$$u = y_m + K_h s y_m$$

u : input of controlled system

s: Laplace operator

When this is assigned to the transfer function of the controlled system, it becomes the following equation.

$$y = (K_h s + 1) / (K s + 1) \cdot y_m$$

5 Here, when error e between the target value y_m and the actual output y is defined as $e = y_m - y$ and the above equation is assigned, the error e is expressed as follows.

$$e = \{1 - (K_h s + 1) / (K s + 1)\} y_m$$

$$= (K - K_h) s / (K s + 1) \cdot y_m$$

10 $= 1 / (K s + 1) \cdot (K - K_h) s y_m$

Because $1 / (K s + 1)$ is a strictly positive real ($K > 0$) in the above equation, the following equation is obtained by adaptive control theory.

$$d(K - K_h) / dt = -\gamma \cdot d y_m / dt \cdot e \quad (\gamma > 0)$$

15 The following equation is derived from the above equation.

$$d K_h / dt = \gamma \cdot d y_m / dt \cdot e$$

Using the above equation, $K_h \rightarrow K$ is guaranteed by adjusting K_h (estimated value of the time constant K).

Thus, by controlling with the above equation using K_h
20 calculated by the above equation, the controlled value y can be made to match the target value y_m .

$$u = y_m + K_h \cdot d y_m / dt$$

From the above equation, an F/F corrected value (feedforward corrected value) u_{cmp} is represented by the following equation.

25 $u_{cmp} = K_h \cdot d y_m / dt$

The ECU 27 periodically executes the electronic throttle control program of Fig. 3, whereby it functions as gain calculating

means and feedforward corrected value calculating means which are referred to in the present invention, the ECU 27 adaptively determines the gain K_h (estimated value of the time constant K) on the basis of a value z obtained by multiplying a derivative value Δy_m of the target throttle opening by the error e between the target throttle opening y_m (target value) and the actual throttle opening y (actual controlled value), and determines, as the F/F corrected value u_{cmp} , a value obtained by multiplying the derivative value Δy_m of the target throttle opening by the gain K_h .

In this case, when the error e between the target throttle opening y_m and the actual throttle opening y is determined with consideration given to the fact that the controlled system has dead time d , a target throttle opening y_{md} at the point in time going back in the past by the amount of the dead time d is used to obtain error $e = y_{md} - y$. The specific processing content of the electronic throttle control program of Fig. 3 will be described below.

When the program is started, first the actual throttle opening y (actual controlled value) is measured in step 101 by the throttle opening sensor 16, and the target throttle opening $y_m(i)$ is calculated in step 102 on the basis of the accelerator opening. Thereafter, the program proceeds to step 103, where the difference value Δy_m (derivative value of the target value) between the current value $y_m(i)$ of the target throttle opening and the previous value $y_m(i - 1)$ is calculated.

$$\Delta y_m = y_m(i) - y_m(i - 1)$$

Then, in step 104, the target throttle opening $y_m(i - d)$ at the point in time going back in the past by the amount of the dead

time d is read and dead time processing is implemented. Thereafter, the program proceeds to step 105, where the error $e (= y_{md} - y)$ between the target throttle opening y_{md} and the actual throttle opening y is calculated.

5 Thereafter, the program proceeds to step 106, where the value $z (= e \times \Delta y_m)$, which is obtained by multiplying the target throttle opening difference value Δy_m by the error e , is calculated. Thereafter, the program proceeds to step 107, where the gain K_h (estimated value of the time constant K) is calculated by the following
10 equation.

$$K_h = K_h (i - 1) + \gamma_k \times \Delta t \times z$$

Here, $K_h (i - 1)$ is the previous gain, γ_k is a constant (> 0) and Δt is the control period.

Then, in step 108, the target throttle opening difference value
15 Δy_m is multiplied by the gain K_h to determine the F/F corrected value u_{cmp} .

Thereafter, the program proceeds to step 109, where another corrected value u_{other} such as an F/B corrected value is calculated. Thereafter, the program proceeds to step 110, where the other
20 corrected value u_{other} is added to the F/F corrected value u_{cmp} to determine the control input u .

$$u = u_{cmp} + u_{other}$$

It should be noted that the program may also be configured so that u_{cmp} and u_{other} are determined by a correction factor and
25 u_{cmp} and u_{other} are multiplied by a base value to determine the control input u .

Then, in step 111, the motor 17 is driven by the control input

u so that the actual throttle opening y is made to match the target throttle opening y_m .

In the above-described first embodiment, the electronic throttle system is configured so that the F/F control is corrected
5 by adaptive control. Thus, the gain K_h of the F/F control can be automatically adjusted in accordance with characteristic variations in the controlled system (electronic throttle system), F/F control reflecting effects resulting from changes in the characteristics of the controlled system can be conducted, and the control precision
10 of the F/F control can be improved. Moreover, because the control equation calculating the input (control input u) of the controlled system from the target throttle opening y_m (target value) serves as an inverse model of the transfer function of the controlled system, the output of the controlled system (actual throttle opening y)
15 can be made to match the target value (target throttle opening y_m), and highly responsive F/F control can be realized.

Moreover, in the first embodiment, when the error e between the target throttle opening y_m (target value) and the actual throttle opening y (actual controlled value) is determined in consideration
20 of the fact that the electronic throttle system, which is the controlled system, has dead time d , the target throttle opening y_{md} at the point in time going back in the past by the amount of the dead time d is used to obtain error $e = y_{md} - y$. Thus, even in a case where the controlled system has dead time d , F/F control
25 where the effects of the dead time d have been removed can be executed, and the control precision of the F/F control can be excellently maintained.

Thus, as shown in Figs. 4A and 4B, in the first embodiment, highly responsive and highly precise electronic throttle control can be realized by correction resulting from adaptive control in comparison to a conventional system where there is no correction
5 resulting from adaptive control.

Second Embodiment

Next, a second embodiment where the invention is applied to an air-fuel ratio control system will be described on the basis of Figs. 5 and 6. When an air-fuel control system is used as the
10 controlled system, consideration is given to the fact that the target value is the target fuel amount and the output (controlled value) of the controlled system becomes the air-fuel ratio (A/F ; excess air ratio λ , excess fuel ratio ϕ) detected by the air-fuel ratio sensor 24 disposed at the exhaust pipe 22, the gain K_h is adaptively
15 determined on the basis of the value z obtained by multiplying the target fuel amount difference value Δy_m (derivative value of the target fuel amount) by the error e between the target excess fuel ratio (represented below as "target ϕ ") and the actual excess fuel ratio (represented below as "actual ϕ ") detected by the air-fuel
20 ratio sensor 24, and a value obtained by multiplying the target fuel amount difference value Δy_m by the gain K_h is determined as the F/F corrected value u_{cmp} . In this case, when the error e between the target ϕ and the actual ϕ is determined in consideration of the fact that the controlled system has dead time d , the target ϕ ($=\phi_d$
25 $=\phi(i-d)$) at the point in time going back in the past by the amount of the dead time d is used to obtain the error $e = \text{target } \phi_d - \text{actual } \phi$. The specific processing content of an air-fuel ratio control

program of Fig. 6 will be described below.

When the program is started, first the intake air amount and the air-fuel ratio are measured in step 201, and the target fuel amount $y_m(i)$ is calculated in step 202 on the basis of the intake
5 air amount. Thereafter, the program proceeds to step 203, where the difference value Δy_m (derivative value of the target value) between the current value $y_m(i)$ of the target fuel amount and the previous value $y_m(i - 1)$ is calculated.

$$\Delta y_m = y_m(i) - y_m(i - 1)$$

10 Then, in step 204, the actual $\phi (= 1 / \lambda)$ is calculated from the measured air-fuel ratio. Thereafter, the program proceeds to step 205, where the target $\phi(i - d)$ at the point in time going back in the past by the amount of dead time d is read and dead time processing, where $\phi_d = \phi(i - d)$, is implemented. Thereafter, the program proceeds
15 to step 206, where the error $e (= \text{target } \phi_d - \text{actual } \phi)$ between the target ϕ_d at the point in time going back in the past by the amount of the dead time d and the actual ϕ is calculated.

Thereafter, the program proceeds to step 207, where the value $z (= e \times \Delta y_m)$, which is obtained by multiplying the target fuel
20 amount difference value Δy_m by the error e , is calculated. Thereafter, the program proceeds to step 208, where the gain K_h (estimated value of the time constant K) is calculated by the following equation.

$$K_h = K_h(i - 1) + \gamma_k \times \Delta t \times z$$

Here, $K_h(i - 1)$ is the previous gain, γ_k is a constant (> 0) and Δt is the control period.
25

Then, in step 209, the target fuel amount difference value Δy_m is multiplied by the gain K_h to determine the F/F corrected

value ucmp.

$$ucmp = K_h \times \Delta y_m$$

Thereafter, the program proceeds to step 210, where another corrected value uother such as a basic injection amount and an F/B
5 corrected value is calculated. Thereafter, the program proceeds to step 211, where the other corrected value uother is added to the F/F corrected value ucmp to determine the control input u.

$$u = ucmp + uother$$

It should be noted that the program may also be configured
10 so that ucmp and uother are determined by a correction factor and ucmp and uother are multiplied by a base value to determine the control input u.

Then, in step 212, the fuel injection valve 20 is driven by the control input u so that the actual ϕ is made to match the target
15 ϕ .

In the above-described second embodiment, the air-fuel ratio control system is configured so that the F/F control is corrected by adaptive control. Thus, highly responsive and highly precise air-fuel ratio control can be realized.

Moreover, in the second embodiment, in consideration of the
20 fact that the target value is the target fuel amount and the output of the controlled system becomes the air-fuel ratio, the excess fuel ratio ϕ , which is the inverse number ($1 / \lambda$) of the excess air ratio, is used as the air-fuel ratio information rather than the
25 excess air ratio λ . Thus, there is the advantage that the directions of increase and decrease of the target value (target fuel amount, target ϕ) and the output of the controlled system (actual ϕ) match

and it becomes easier to understand the behavior of the controlled system.

Third Embodiment

A third embodiment where the invention is applied to an electronic throttle system will be described on the basis of Figs. 7 and 8.

In the first and second embodiments, a controlled system was approximated by a first-order lag system, but in the third embodiment, the controlled system is approximated as shown in Fig. 7 in order to more accurately model the controlled system. In this case, the output y (actual throttle opening) of the controlled system can be made to match the target value y_m (target throttle opening) as long as the control (inverse model of the transfer function of the controlled system) in the dotted lines of Fig. 7 can be realized.

However, because constants K_1 and K_2 of the controlled system are unknown or vary, they cannot be expressed with the control equation of Fig. 7.

Thus, in the third embodiment, a method that detects and controls the constants K_1 and K_2 of the controlled system is adopted.

First, the transfer function of the control equation $K_1 K_2 s / (K_1 s + 1)$ is transformed to make it easy to develop.

$$\begin{aligned} K_1 K_2 s / (K_1 s + 1) &= K_2 s / (s + 1/K_1) \\ &= \beta s / (s + \alpha) \end{aligned}$$

Here, $\alpha = 1 / K_1$ and $\beta = K_2$.

Moreover, the transfer function of the control equation $(K_1 s + 1) / \{K_1 (1 + K_2) s + 1\}$ is transformed to make it easy to develop.

$$\begin{aligned}
& (K_1 s + 1) / \{K_1 (1 + K_2) s + 1\} \\
& = (s + 1 / K_1) / \{(1 + K_2) s + 1 / K_1\} \\
& = (s + \alpha) / \{(1 + \beta) s + \alpha\}
\end{aligned}$$

Thus, the relation between the input u (control input) and
5 the output y (controlled value) of the controlled system is
represented by the following equation.

$$y = (s + \alpha) / \{(1 + \beta) s + \alpha\} \cdot u \quad [1]$$

Also, the relation between the target value y_m and the control
input u is represented by the following equation.

$$10 \quad u = \{1 + \beta h s / (s + \alpha h)\} y_m \quad [2]$$

Here, αh represents the estimated value of α and βh represents
the estimated value of β .

When equation [2] is assigned to equation [1], it becomes as
follows.

$$\begin{aligned}
15 \quad y &= (s + \alpha) / \{(1 + \beta) s + \alpha\} \cdot \{1 + \beta h s / (s + \alpha h)\} y_m \\
&= (s + \alpha) / \{(1 + \beta) s + \alpha\} \\
&\quad \times \{(1 + \beta h) s + \alpha h\} / (s + \alpha h) \times y_m
\end{aligned}$$

Here, when the error e between the target value y_m and the
actual output y is defined as $e = y_m - y$ and the above equation
20 is assigned, the error e is represented as follows.

$$\begin{aligned}
25 \quad e &= \frac{(\beta - \beta h) s^2 + (\alpha h \beta - \alpha \beta h) s}{(1 + \beta) s^2 + \{(1 + \beta) + \alpha h + \alpha\} s + \alpha \alpha h} \times y_m \\
&= \frac{s}{(1 + \beta) s^2 + \{(1 + \beta) + \alpha h + \alpha\} s + \alpha \alpha h} \times [\beta - \beta h (\alpha h \beta - \alpha \beta h)] \\
30 \quad &\times \left(\frac{dy_m}{dt} \right)_{y_m}
\end{aligned}$$

$$\epsilon = e + c_1 \int e dt$$

$$\begin{aligned} 5 \quad \epsilon &= \frac{s + c_1}{(1 + \beta) s^2 + \{(1 + \beta) \alpha h + \alpha\} s + \alpha \alpha h} \times [\beta - \beta h (\alpha h \beta - \alpha \beta h)] \\ 10 \quad &\times \left(\frac{dym}{dt} \right)_{ym} \end{aligned}$$

$$15 \quad \frac{d\phi_1}{dt} = - \gamma_\beta \frac{dym}{dt} \epsilon \quad (\phi_1 = \beta - \beta h, \phi_2 = \alpha h \beta - \alpha \beta h)$$

$$\frac{d\beta h}{dt} = \gamma_\beta \frac{dym}{dt} \epsilon \quad [3]$$

20 When βh becomes β , ϵ is represented as follows.

$$\begin{aligned} 25 \quad \epsilon &= \frac{\beta (s + c_1)}{(1 + \beta) s^2 + \{(1 + \beta) \alpha h + \alpha\} s + \alpha h} \begin{pmatrix} 0 & \alpha h - \alpha \end{pmatrix} \\ &\times \left(\frac{dym}{dt} \right)_{ym} \end{aligned}$$

$$30 \quad \frac{d\phi_2'}{dt} = - \gamma_\alpha ym \epsilon \quad (\phi_2' = \alpha h - \alpha)$$

$$35 \quad \frac{d\alpha h}{dt} = - \gamma_\alpha ym \epsilon \quad [4]$$

40 When the F/F corrected value $ucmp$ is expressed as an equation,
it becomes the following.

$$\begin{aligned} ucmp &= K2h \cdot ym - 1 / K1h \cdot \int ucmp \cdot dt \\ &= \beta h \cdot ym - \alpha h \cdot \int ucmp \cdot dt \end{aligned}$$

Here, βh and αh are determined from the relation of equation [3] and equation [4].

In this case, if the error ϵ is not 0, the da_h / dt of equation [4] does not become 0 and the problem arises that αh continues to be updated. In other words, if it has a steady-state deviation, the problem arises that αh always continues to be updated.

Thus, in the third embodiment, in order to update αh to only the scene where the F/F control works, the following equation, where the previous F/F corrected value $ucmp$ is multiplied by the right part of equation [4], is used to calculate αh .

$$da_h / dt = - \gamma \alpha \cdot y_m \cdot \epsilon \cdot ucmp$$

$$= - \gamma \alpha \cdot z_1$$

$$z_1 = y_m \cdot \epsilon \cdot ucmp$$

The ECU 27 periodically executes the electronic throttle control program of Fig. 8, whereby it functions as gain calculating means and feedforward corrected value calculating means which are referred to in the scope of the patent claims, the ECU 27 adaptively determines the gain $K2h$ on the basis of a value z_2 obtained by multiplying the derivative value Δy_m of the target throttle opening by the sum $(e + c \cdot \int edt)$ of the error e between the target throttle opening y_m (target value) and the actual throttle opening y (actual controlled value) and the integral value of that error, and determines, as the F/F corrected value $ucmp$, a value obtained by multiplying the gain $K2h$ by the difference value $(y_m - u_1)$ between the target throttle opening y_m and a value u_1 of the first-order lag of the target throttle opening.

In this case, when the value u_1 of the first-order lag of the

target throttle opening y_m is calculated, the first-order lag time constant (estimated value of K_1) thereof is adaptively determined on the basis of the value z_1 obtained by multiplying the target throttle opening y_m and the previous F/F corrected value $ucmp$ by the sum
5 $\epsilon (= e + ee)$ of the error e between the target throttle opening y_m and the actual throttle opening y and the integral value ee of that error.

Moreover, when the error e between the target throttle opening y_m and the actual throttle opening y is determined in consideration
10 of the fact that the controlled system has dead time d , the target throttle opening y_{md} at the point in time going back in the past by the amount of the dead time d is used to obtain error $e = y_{md} - y$. The specific processing content of the electronic throttle control program of Fig. 8 will be described below.

15 When the program is started, first the actual throttle opening y (actual controlled value) is measured in step 301 by the throttle opening sensor 16, and the target throttle opening $y_m(i)$ that is the target value is calculated in step 302 on the basis of the accelerator opening. Thereafter, the program proceeds to step 303,
20 where the difference value Δy_m (derivative value of the target value) between the current value $y_m(i)$ of the target throttle opening and the previous value $y_m(i - 1)$ is calculated.

$$\Delta y_m = y_m(i) - y_m(i - 1)$$

Then, in step 304, the target throttle opening $y_m(i - d)$ at
25 the point in time going back in the past by the amount of the dead time d is read and dead time processing is implemented. Thereafter, the program proceeds to step 305, where the error $e (= y_{md} - y)$

between the target throttle opening y_{md} and the actual throttle opening y is calculated.

Thereafter, the program proceeds to step 306, where the integral value ee of the error e is calculated by the following
5 equation.

$$ee = ee + c \times \Delta t \times e$$

(c : constant; Δt : control period)

Then, in step 307, the sum ε ($= e + ee$) of the error e and the integral value ee thereof is calculated. Thereafter, the program
10 proceeds to step 308, where the target throttle opening y_m and the previous F/F corrected value $ucmp$ are multiplied by ε to determine z_1 .

$$z_1 = \varepsilon \times y_m \times ucmp$$

Thereafter, the program proceeds to step 309, where αh is
15 calculated by the following equation.

$$\alpha h = \alpha h - \gamma \alpha \times \Delta t \times z_1$$

($\gamma \alpha$: constant)

Thereafter, the program proceeds to step 310, where the first-order lag time constant $K1h$ used in calculating the value
20 u_1 of the first-order lag of the target throttle opening y_m is calculated by the following equation using αh .

$$K1h = 1 / \alpha h$$

Then, in step 311, the value z_2 ($= \varepsilon \times \Delta y_m$), which is obtained by multiplying the target throttle opening difference value Δy_m
25 by ε , is calculated. Thereafter, the program proceeds to 312, where the gain $K2h$ (estimated value of the constant K_2) is calculated by the following equation.

$$K2h = K2h (i - 1) + \gamma_2 \times z_2$$

Here, $K2h (i - 1)$ represents the previous gain and γ_2 represents a constant (> 0).

Thereafter, the program proceeds to step 313, where the value u_1 of the first-order lag of the target throttle opening y_m is calculated by the following equation using the first-order lag time constant $K1h$.

$$u_1 = K1h / (K1h + \Delta t) \cdot u_1 + \Delta t / (K1h + \Delta t) \cdot y_m$$

Then, the program proceeds to step 314, where the value obtained by multiplying the gain $K2h$ by the difference value $(y_m - u_1)$ between the target throttle opening y_m and the value u_1 of the first-order lag of the target throttle opening y_m is determined as the F/F corrected value $ucmp$.

$$ucmp = (y_m - u_1) \times K2h$$

Thereafter, the program proceeds to step 315, another corrected value u_{other} such as an F/B corrected value is calculated. Thereafter, the program proceeds to step 316, where the other corrected value u_{other} is added to the F/F corrected value $ucmp$ to determine the control input u .

$$u = ucmp + u_{other}$$

It should be noted that the program may also be configured so that $ucmp$ and u_{other} are determined by a correction factor and $ucmp$ and u_{other} are multiplied by a base value to determine the control input u .

Then, in step 317, the motor 17 is driven by the control input u so that the actual throttle opening y is made to match the target throttle opening y_m .

In the electronic throttle control of the above-described third embodiment, control precision can be further improved over the first embodiment because the precision of the model of the controlled system is improved over the first embodiment.

5 *Fourth Embodiment*

Next, a fourth embodiment where the invention is applied to an air-fuel ratio control system will be described on the basis of Figs. 9 and 10. Similar to the second embodiment, when an air-fuel control system is used as the controlled system, consideration is
10 given to the fact that the output y (air-fuel ratio) of the controlled system is detected by the air-fuel ratio sensor 24 disposed at the exhaust pipe 22, the gain $K2h$ is adaptively determined on the basis of the value z_2 obtained by multiplying the derivative value Δy_m of the target fuel amount by the sum $(e + c \cdot \int e dt)$ of the error
15 e between the target ϕ (target excess fuel ratio) and the actual ϕ detected by the air-fuel ratio sensor 24 and the integral value of that error, and a value obtained by multiplying the gain $K2h$ by the difference value $(y_m - u_1)$ between the target fuel amount y_m and the value u_1 of the first-order lag of the target fuel amount
20 is determined as the F/F corrected value u_{cmp} . In this case, when the error e between the target ϕ and the actual ϕ is determined in consideration of the fact that the controlled system has dead time d , the target ϕ ($=\phi_d = \phi(i - d)$) at the point in time going back in the past by the amount of the dead time d is used to obtain the
25 error $e = \text{target } \phi_d - \text{actual } \phi$. Also, in order to more precisely model the controlled system, the controlled system is modeled by a commonly known fuel behavior model as shown in Fig. 7. The specific

processing content of the air-fuel ratio control program of Fig. 9 will be described below.

When the program is started, first the intake air amount and the air-fuel ratio are measured in step 401, and the target fuel amount $y_m(i)$ is calculated in step 402 on the basis of the intake air amount. Thereafter, the program proceeds to step 403, where the difference value Δy_m (derivative value of the target value) between the current value $y_m(i)$ of the target fuel amount and the previous value $y_m(i-1)$ is calculated.

10
$$\Delta y_m = y_m(i) - y_m(i-1)$$

Then, in step 404, the actual $\phi (= 1 / \lambda)$ is calculated from the measured air-fuel ratio. Thereafter, the program proceeds to step 405, where the target $\phi(i-d)$ at the point in time going back in the past by the amount of dead time d is read and dead time processing, where $\phi_d = \phi(i-d)$, is implemented. Thereafter, the program proceeds to step 406, where the error $e (= \text{target } \phi_d - \text{actual } \phi)$ between the target ϕ_d at the point in time going back in the past by the amount of the dead time d and the actual ϕ is calculated.

Thereafter, the program proceeds to step 407, where the integral value ee of the error e is calculated by the following equation.

$$ee = ee + c \times \Delta t \times e$$

(c : constant; Δt : control period)

Then, in step 408, the sum $\varepsilon (= e + ee)$ of the error e and the integral value ee thereof is calculated. Thereafter, the program proceeds to step 409, where the target fuel amount y_m and the previous F/F corrected value u_{cmp} are multiplied by ε to determine z_1 .

$$z_1 = \varepsilon \times y_m \times u_{cmp}$$

Thereafter, the program proceeds to step 410, where α_h is calculated by the following equation.

$$\alpha_h = \alpha_h - \gamma_\alpha \times \Delta t \times z_1$$

5 (γ_α : constant)

Thereafter, the program proceeds to step 411, where the first-order lag constant K_{1h} used in calculating the value u_1 of the first-order lag of the target fuel amount y_m is calculated by the following equation using α_h .

10 $K_{1h} = 1 / \alpha_h$

Then, in step 412, the value $z_2 (= \varepsilon \times \Delta y_m)$, which is obtained by multiplying the target fuel amount difference value Δy_m by ε , is calculated. Thereafter, the program proceeds to 413, where the gain K_{2h} (estimated value of the constant K_2) is calculated by the

15 following equation.

$$K_{2h} = K_{2h} (i - 1) + \gamma_2 \times z_2$$

Here, $K_{2h} (i - 1)$ represents the previous gain and γ_2 represents a constant (> 0).

Thereafter, the program proceeds to step 414, where the value

20 u_1 of the first-order lag of the target fuel amount y_m is calculated by the following equation using the first-order lag time constant K_{1h} .

$$u_1 = K_{1h} / (K_{1h} + \Delta t) \cdot u_1 + \Delta t / (K_{1h} + \Delta t) \cdot y_m$$

Then, in step 415, a value obtained by multiplying the gain

25 K_{2h} by the difference value $(y_m - u_1)$ between the target fuel amount y_m and the value u_1 of the first-order lag of the target fuel amount is determined as the F/F corrected value u_{cmp} .

$$ucmp = (y_m - u_1) \times K_2h$$

Thereafter, the program proceeds to step 416, where another corrected value u_{other} such as a basic injection amount and an F/B corrected value is calculated. Thereafter, the program proceeds
 5 to step 417, where the other corrected value u_{other} is added to the F/F corrected value $ucmp$ to determine the control input u .

$$u = ucmp + u_{other}$$

It should be noted that the program may also be configured so that $ucmp$ and u_{other} are determined by a correction factor and
 10 $ucmp$ and u_{other} are multiplied by a base value to determine the control input u .

Then, in step 418, the fuel injection valve 20 is driven by the control input u so that the actual ϕ is made to match the target ϕ .

15 In the air-fuel ratio control of the above-described fourth embodiment, control precision can be further improved over the second embodiment because the precision of the model of the controlled system is improved over the second embodiment.

Figs. 10A and 10B show the behavior of the air-fuel ratio control
 20 of the fourth embodiment. Because the fourth embodiment is configured so that F/F control is corrected by adaptive control, variations in the actual ϕ of the transient state can be effectively reduced by the F/F corrected value $ucmp$ resulting from adaptive control, and driveability in the transient state and exhaust
 25 emissions can be improved.

Fifth Embodiment

In equations [3] and [4] of expression [1] described in the

third embodiment, ε (sum of the error e between the target value and the actual controlled value and the integral value $e\int$ of that error) was used, but in the fifth embodiment, the error e between the target value and the actual controlled value is used in place
 5 of ε and equations [3] and [4] of expression [1] are changed to the following equations [3'] and [4'].

$$d\beta h / dt = - \gamma \beta \cdot dym / dt \cdot e \quad [3']$$

$$dah / dt = - \gamma \alpha \cdot ym \cdot e \quad [4']$$

In the fifth embodiment also, in order to update ah to only
 10 the scene where the F/F control works, the following equation, where the previous F/F corrected value $ucmp$ is multiplied by the right part of equation [4'], is used to calculate ah .

$$dah / dt = - \gamma \alpha \cdot ym \cdot e \cdot ucmp$$

$$= - \gamma \alpha \cdot z_1$$

$$15 \quad z_1 = ym \cdot e \cdot ucmp$$

In other words, the fifth embodiment uses "error e " in place of " ε " in the third embodiment.

Effects that are the same as those of the third embodiment can be obtained even if the invention is configured in this manner.

20 It should be noted that the range of application of the invention is not limited to an electronic throttle system and an air-fuel ratio control system. The invention can also be applied to and implemented in various control systems disposed in vehicles, such as idle speed control, valve control and cruise control systems.